

Realization of a Photovoltaic Fed Sparse Alternating Current (AC)-Link Inverter

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Abstract: In this paper, a soft-switched alternating current (AC)-link buck-boost inverter with a reduced number of switches, referred to as a sparse AC-link buck-boost inverter, was designed and implemented for a photovoltaic (PV) interface. Important features of the sparse configuration included a lower number of switches, lower failure rates, compactness, and cost-effectiveness. The link was composed of a low reactive rating series inductor/capacitor pair. Significant merits of the AC-link buck-boost inverter are a zero voltage turn on and a soft turn off of the switches, resulting in minimum voltage stress on the switches and negligible switching losses. In this paper, 10 switches were used instead of 20 switches as are used in existing buck-boost inverter topology. The reduction in the number of switches did not change the principle of operation of the sparse configuration; hence, it remains the same as that of the original configuration. The pulse width modulation (PWM) technique was used for gating the switches. The inverter operation was validated and implemented for PV interface using a microcontroller.

Keywords: Sparse inverter, Photovoltaic systems, Zero voltage switching, MATLAB, Microcontroller.

تحقيق التغذية الكهروضوئية عن طريق الربط بمعاكس متفرق للتيار المتردد

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الملخص: في هذه المقالة، تم تصميم وبناء وسيط ربط للخلية الكهروضوئية من خلال تكوينات متفرقة لمعاكس (الإضعاف – التعزيز) باستخدام التحويل الناعم للتيار المتردد باستخدام عدد أقل من المفاتيح. إن من السمات الهامة للتكوينات المتفرقة أنها تحتوي على عدد أقل من المفاتيح وانخفاض معدل الأعطال ومدمجة التكوين بأقل تكلفة. يتألف الربط من زوج ملف حثي / سعوي بشكل متتالي بحيث تكون القيمة الغير فعالة قليلة. هذا المعاكس له مزايا بارزة مثل صغر الجهد عند الإشعال والإطفاء الناعم الذي يؤدي إلى إجهاد قليل للفولت وكمية مفقودات لا تذكر عند التبديل. في هذه المقالة تم استخدام ١٠ مفاتيح بدلا من ٢٠ مفتاح كما هو مستخدم في التكوينات الحالية. تخفيض عدد المفاتيح لا يؤثر على مبدأ التشغيل للتكوينات المتفرقة ومن ثم سيبقى المبدأ كما هو في التكوين الأصلي. تم استخدام تقنية تغيير عرض النبضة لإشعال المفاتيح. تم التحقيق من صحة تشغيل وتنفيذ المعاكس للربط بالخلايا الكهروضوئية باستخدام متحكم ميكرو.

الكلمات المفتاحية: معاكس متفرق، أنظمة كهروضوئية، التبديل من الجهد صفر، متحكم ميكرو

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1. Introduction

Renewable energy is defined as energy from a source that is not depleted when used (Carlson, 1989), and photovoltaic (PV) energy is the most important renewable energy source available. (Shen 2008). PV systems are the fastest growing energy source worldwide because of an increase in energy demand, and the fact that fossil energy sources are finite and expensive. PV energy is renewable, inexhaustible, and non-polluting, and can be used for various applications (Zahedi 2002). Unlike other non-renewable energy sources, solar energy sources produce few or no waste products such as carbon dioxide and other harmful chemical substances (Subudhi and Pradhan 2013).

The major challenge in PV conversion is to use the proper power converter circuits in order to convert produced direct current (DC) to practical DC-alternating current (AC) (Chakraborty *et al.* 2009). In a standalone or grid-connected system, the conversion of DC to AC with the proper power quality is important (Atcitty *et al.* 2011). A number of PV inverter topologies have been presented in the literature (Kerekes *et al.* 2011). The major challenges are a reduction of switches to reduce losses and the generation of nearly sinusoidal voltage with reduced total harmonic distortion (Kjaer *et al.* 2005). These converters, which can be configured in many ways, including DC-DC, DC-AC, AC-DC, and AC-AC and have many advantages over other types of converters. Specifically, these converters are more compact and reliable, and offer longer life-time (Qin *et al.* 2002). However, one complication is that these converters employ more switches than any other power processors (Sood *et al.* 1988). Their remarkable merits have paved the way for a soft-switching AC-link universal power converter which has received noticeable attention during the last few years (Amirabadi *et al.* 2009).

In spite of advantages over the AC-link buck-boost inverter, a soft-switching AC-link universal power converter requires many switches when compared to a more conventional inverter (Amirabadi 2013); hence, the control method is complicated. In a conventional AC-link buck-boost inverter, the complementary switches are added to provide the pathway for an AC-link current

(Balakrishnan *et al.* 2008). This AC-link current may reduce the system efficiency with an increased harmonic content. Due to this potential issue, the sparse AC-link inverter is considered, and the configuration with the reduced number of switches is called a sparse AC-link buck-boost inverter. The operating principle of a sparse AC-link inverter is similar to that of a conventional one. The proposed inverter has high reliability and low switching loss (Toliat *et al.* 2008). The most important feature of a sparse inverter is that it has unidirectional switches which can be fabricated by switch modules. The sparse configuration is considered to be less expensive, less complicated, and more compact compared to the AC-link buck-boost inverter (Amirabadi 2013). The sparse AC-link inverter supports bidirectional power flow. In this inverter, the complimentary switches are not essential, although they provide AC (Amirabadi 2014). The converter transfer power through the DC-link inductor is charged in the input phase and discharged at the output phase. The link frequency is greater than the output line frequency. During the resonance mode, no switches conduct, initiating the soft switching principle (Amirabadi *et al.* 2009). However, due to the number of switches in the sparse AC-link inverter, the switching loss can be increased, affecting the system's efficiency. Hence, in this paper, the sparse AC-link inverter is proposed with a reduced number of switches. Due to the reduced number of switches, the control method is very simple compared to a conventional AC-link inverter. In the proposed method, a reduced number of switches is used which makes the control process easier. A soft-switching AC-link buck-boost inverter traditionally uses 20 switches, but the sparse configuration proposed by Amirabadi *et al.* (2014) used 18 switches instead of 20.

In this paper, a sparse configuration with 10 switches was implemented for PV interface. The operation was checked with MATLAB/SIMULINK software (MathWorks, Natick, Massachusetts, USA) and validated through practical implementation. The paper is organized as follows: Section 2 describes the design and operation of sparse configuration. Section 3 presents the simulation results, and section 4 presents the hardware implementation. The conclusion is presented in section 5.

2. Design and Operation of PV Connected Sparse AC-link Inverter

The basic circuit configuration of PV connected sparse AC link inverter is shown in Fig 1. The input side and the output side are formed by four unidirectional switches and six bi-directional switches respectively. This inverter is most similar to a DC-AC buck-boost converter in which the link is charged through PV array and discharged into the output phases.

The charging and discharging of the link in reverse is facilitated by the complementary switches in each leg. This style of discharging leads to an AC in the link, which leads to better utilization of the link inductor and capacitor.

A single-phase high-frequency transformer can be added to the link if isolation is required. The PV panel was modelled using a one-diode model which consists of a current source in parallel with a diode (Villava *et al.* 2009), a shunt resistance, and a series resistance (Fig. 2). Key modelling is shown in Eqns. 1-4.

In this case, the link inductance of the transformer is playing the role of magnetizing inductance (Amirabadi *et al.* 2014).

$$I_o = \frac{I_{scn} + K_i dT}{\exp\left[\frac{V_{ocn} + K_v dT}{V_{ta}}\right] - 1} \quad (1)$$

$$I_{pv} = [K_i dT + I_{pvn}] \frac{G}{G_n} \quad (2)$$

$$V_{ta} = \frac{N_s a K T}{q} \quad (3)$$

$$I_m = I_{pv} N_{pp} - I_o N_{pp} \quad (4)$$

$$\exp\left[\frac{V_{ta}}{N_s} \times \left(R_s \frac{N_{ss}}{N_{pp}} I + V\right) - 1\right]$$

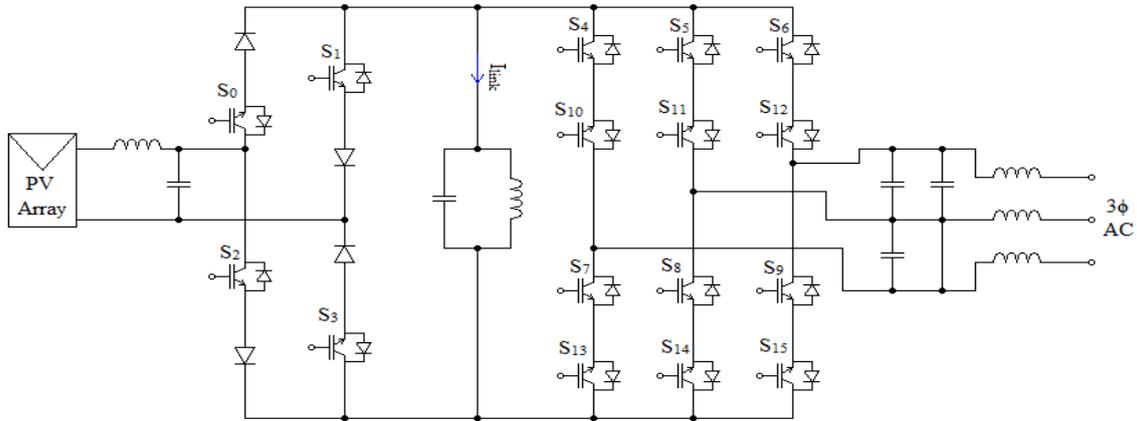


Figure 1. Photovoltaic (PV) interfaced sparse alternating current (AC)-link inverter.

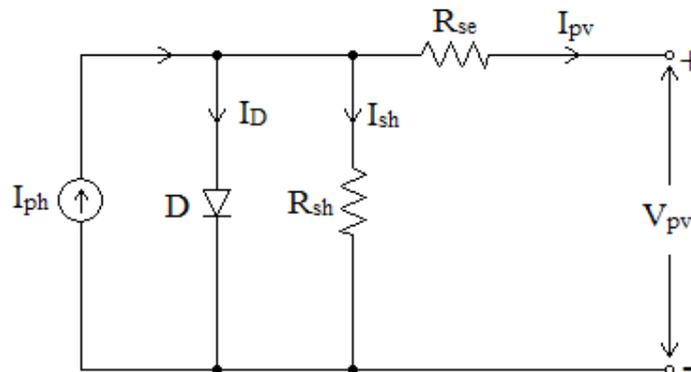


Figure 2. Circuit model used for modelling a photovoltaic (PV) panel.

The PV array is sized accordingly (considering maximum power point voltage, V_{mpp} and current, I_{mpp}) and the system parameters are listed in Table 1.

Table 1. System parameters.

Parameters	Standards
V_{mpp} of single panel	16.54 V
I_{mpp} of single panel	2.55 A
PV array size/arm	15×2 (15 panels in series)

PV = photovoltaic; V = volts; A = amps.

The design of an AC link PV inverter is explained below (Amirabadi *et al.* 2012). For the purpose of simplicity, a resonating time which is less than the power transfer time at full power is neglected. The phase carrying maximum current is involved in de-energizing of the link at modes three and five. The equivalent output voltage and current is given by

$$V_{oeq} = \frac{\pi V_{opeak}}{2} \cos\theta \quad (5)$$

$$I_{oeq} = \frac{3I_{opeak}}{\pi} \quad (6)$$

where, I_{opeak} is the output peak phase current, V_{opeak} is the output peak phase voltage, and $\cos(\theta)$ is the output power factor. The link current during energizing and de-energizing is given by

$$I_{link,peak} = \frac{V_{pv} \times t_{charge}}{L} \quad (7)$$

$$I_{link,peak} = \frac{V_{oeq} \times t_{discharge}}{L} \quad (8)$$

where, t_{charge} and $t_{discharge}$ are the total charging times during mode one and total discharging time during modes three and five.

The average PV current based on the peak of the link current is given as follows:

$$I_{pv} = \frac{1}{T} \left(I_{link,peak} \times t_{charge} \right) \quad (9)$$

where, T is the period of the link current. Based on the above equations, the peak link current can be written as:

$$I_{link,peak} = 2 \times (I_{pv} + I_{oeq}) \quad (10)$$

The frequency of the link current is determined based on the power rating of the system and the characteristics of available switches. The link inductance is calculated by:

$$L = \frac{V_{pv}}{I_{link,peak}} \frac{1}{2f \left(1 + \left(\frac{V_{pv}}{V_{oeq}} \right) \right)} \quad (11)$$

Link capacitance is calculated such that the resonating periods within 5% of the link cycle at full power. There are 12 operating modes in 10 switch sparse AC link inverter. The link is energized from the input during modes 1 and 7 and is de-energized to the outputs during modes 3, 5, 9 and 11. Modes 2, 4, 6, 8, 10 and 12 are devoted to resonance. The link current during mode 1 is given by as below:

$$V_{pv} = L \frac{di_{link}(t)}{dt} \quad (12)$$

$$i_{Link}(t) = \frac{1}{L} \int_0^t V_{pv} dt = \frac{V_{pv} \times t}{L} + i_{Link}(0) \quad (13)$$

The operation in different modes is depicted in Table 2. The operating principle is shown in Fig. 3 for each mode (Amirabadi 2013).

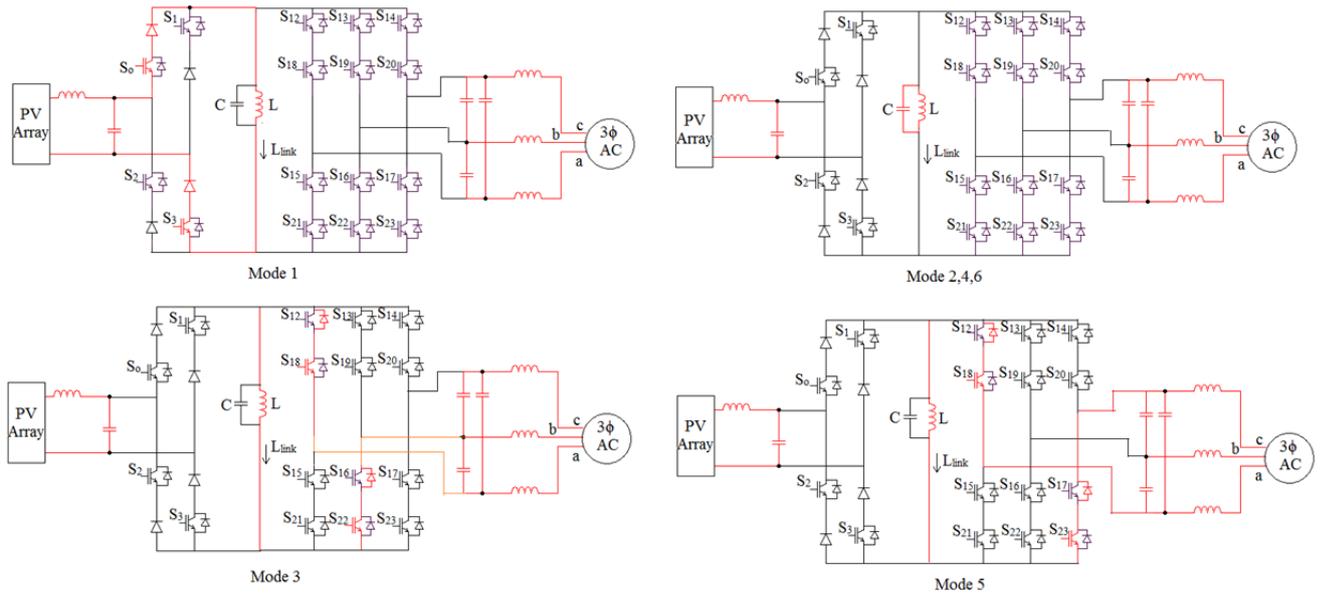


Figure 3. Operating principle of sparse AC link inverter.

Table 2. Modes of operation of 10 switch sparse AC-link inverter.

Mode	Link operation	Switches in ON state	Switches in conduction
1	Energizing	S_0 and S_3	S_0 and S_3
2	Partial Resonance	S_{10} , S_{14} and S_{15}	None
3	De-energizing	S_{10} , S_{14} and S_{15}	S_{10} and S_{14}
4	Partial Resonance	S_{10} and S_{15}	None
5	De-energizing	S_{10} and S_{15}	S_{10} and S_{15}
6	Partial Resonance	S_1 and S_2	None
7-12	The operations are similar to modes 1 to 6. But the link charges and discharges in the reverse direction through complimentary switch at each leg.		

Table 3. Simulation Parameters

Parameter	Values
Nominal PV voltage (DC link)	240 V
Output voltage	230 V
Link inductance	5.5 Mh
Link capacitor	1000 Mf
PV side filter	0.001 μ H, 1000 μ F
AC side filter	0.200 μ H, 200 μ F

PV = photovoltaic; DC = direct current; V = volt; Mh = megahertz; Mf = megaXXX; AC = alternating current; Mh = millihenry; Mf = millifarad.

3. Simulation Results

The circuit shown in Fig. 1 was simulated for a 1kW PV system using MATLAB/SIMULINK (Fig. 4), and the simulation parameters are

presented in Table 3. The filter components were designed using the equations given by Amirabadi *et al.* (2014). Pulse width modulation (PWM) pulses were generated in proper sequence to turn on the switches.

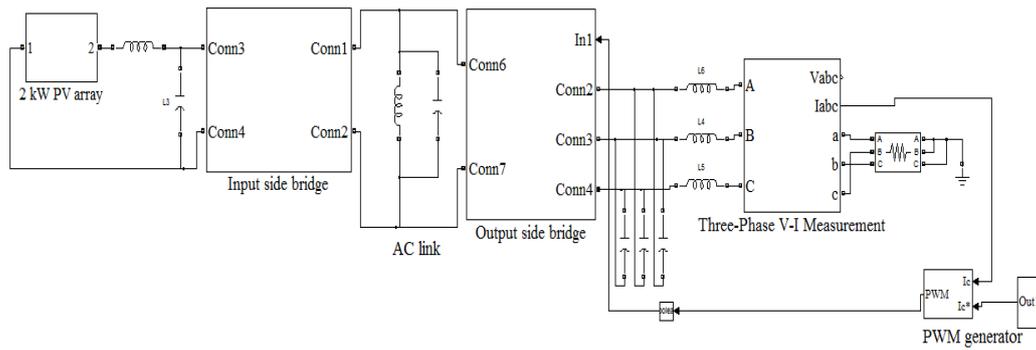


Figure 4. MATLAB model of the 10-switch sparse alternating current (AC)-link inverter.

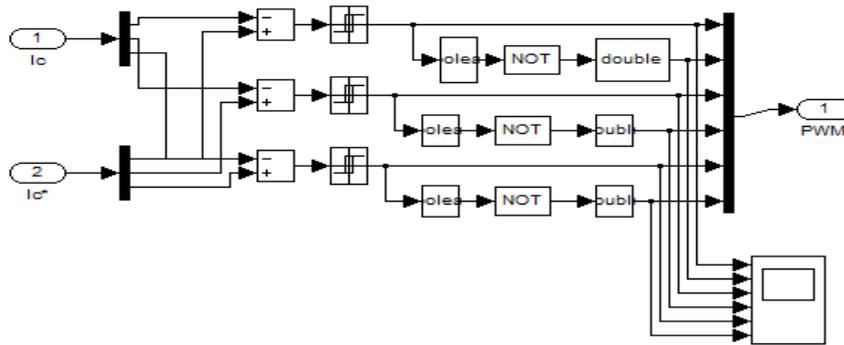


Figure 5. Control circuit for pulse width modulation (PWM) generation.

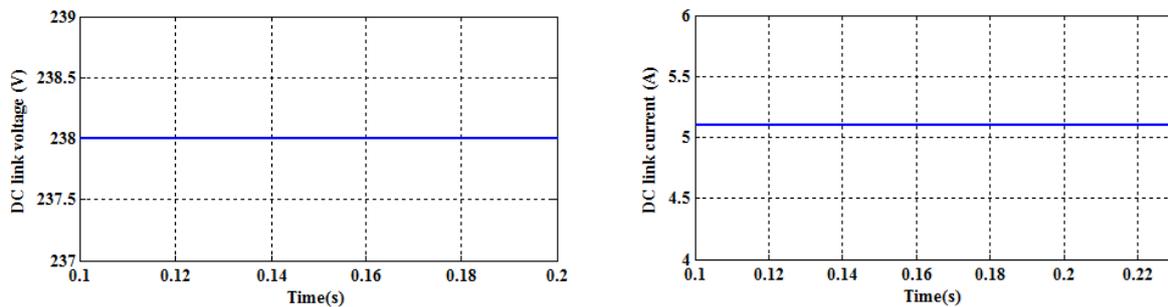


Figure 6. Voltage and current at the direct current (DC) link.

The nominal voltage of 240 V forming the PV array is given as an input-to-input side bridge, while the output is observed from the three legs of the output side bridge. The input side bridge consists of unidirectional switches while the output side consists of bidirectional switches. The PWM pulses are generated through the control circuit whereas the output current is compared with the reference current (Fig. 5).

The simulation was carried out using MATLAB, and the results are presented below. Figure 6 shows the DC voltage and current from The PV array. The maximum power from the PV array was tracked by the maximum power

point tracking algorithm. The maximum power was tracked from the PV array only when the source impedance matched the load impedance.

Figure 7 shows the PWM pulses for switches S_0 – S_3 . The gating sequence to switches S_4 – S_{15} is given by comparing the output current from the inverter (Fig. 8).

Figure 9 shows the input from the DC link voltage and current. In such a set-up, when the capacitor starts charging, the current flowing through the system decreases; when it starts discharging, the current flowing through the system increases. The frequency of the link current is much higher than the frequency of the line.

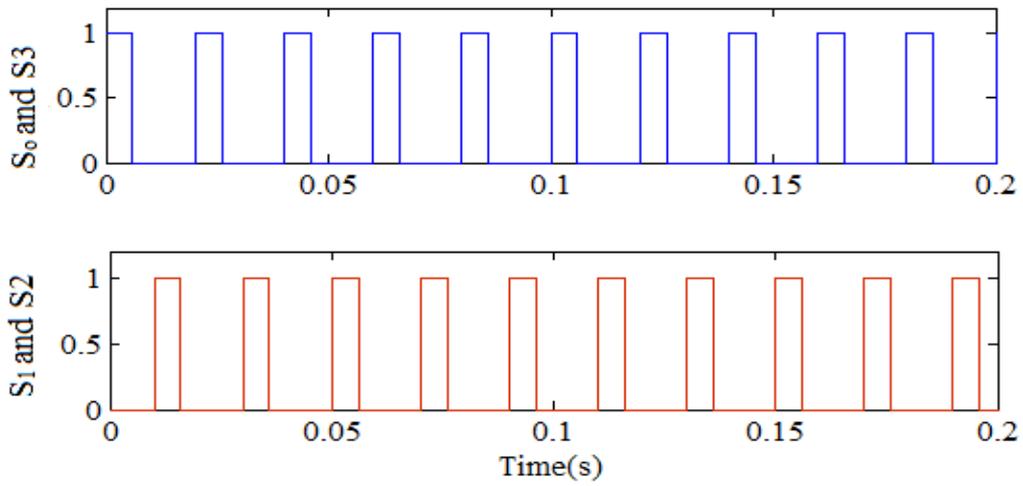


Figure 7. Pulses for the switches in the input side bridge.

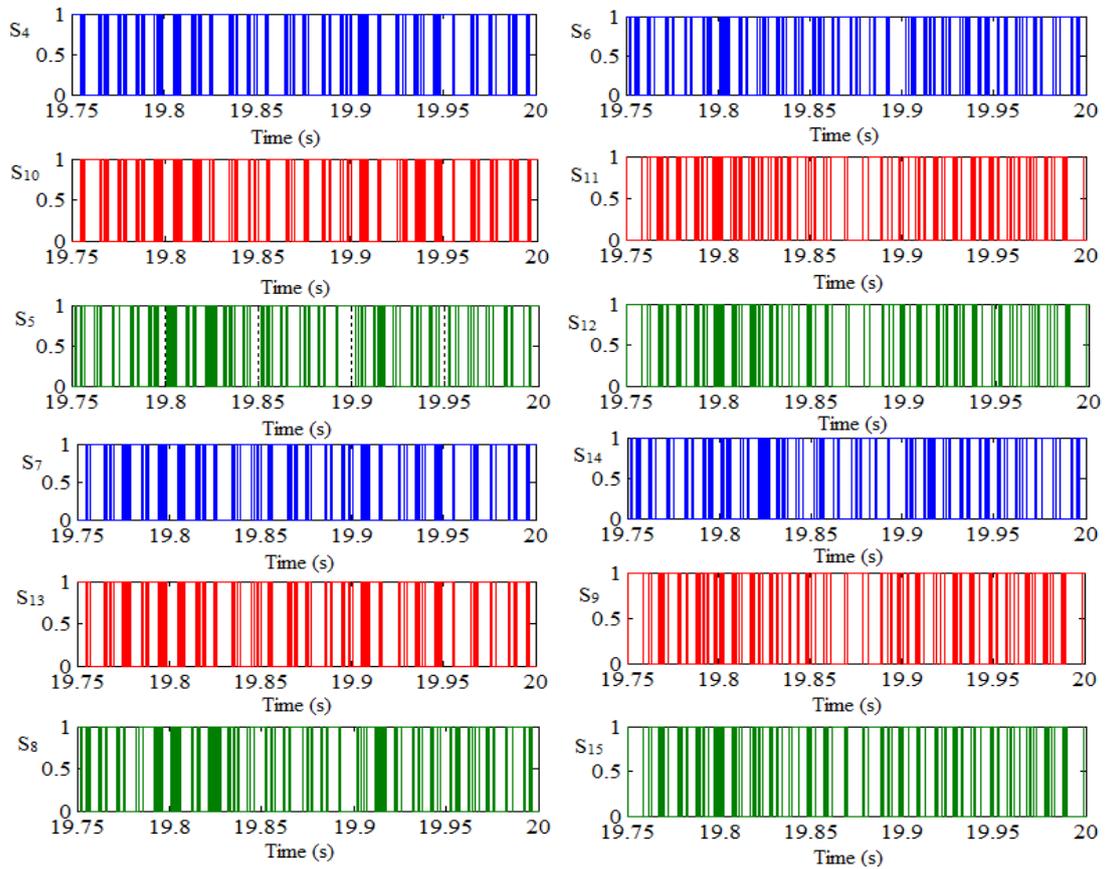


Figure 8. Pulses for the switches in the output side bridge.

Soft switching is carried out in the sparse AC link inverter. Accordingly, there is zero voltage switching when the voltage across the switch is zero. Then the capacitor starts charging. In case

of zero current switching, when the voltage across the switch is zero, the current flowing through the inductor is also zero. The soft switching principle is shown in Fig. 10

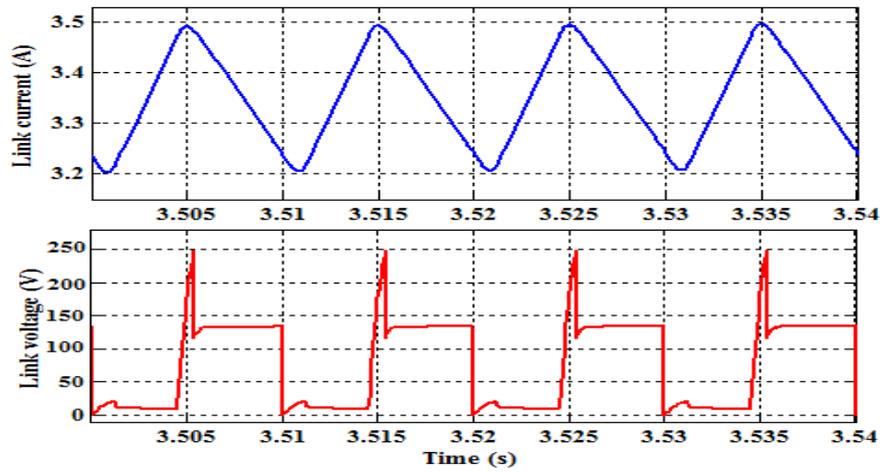


Figure 9. Input dc link current and voltage.

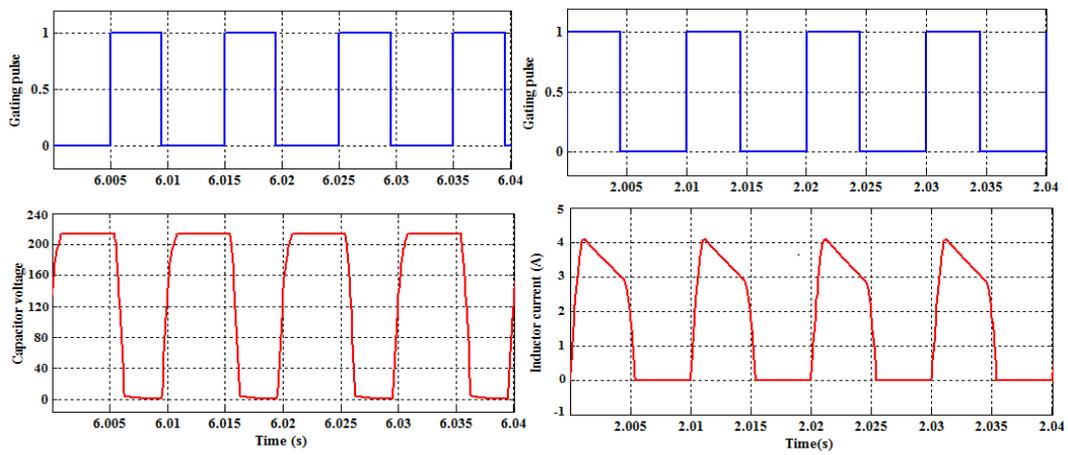


Figure 10. Zero voltage and zero current switching.

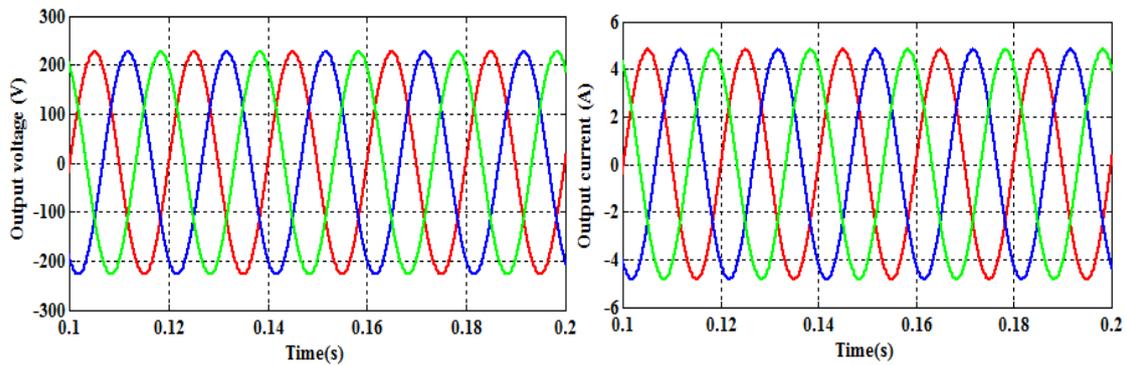


Figure 11. Voltage and current at the alternating current (AC) side.

Figure 11 shows the output voltage and current after filter. The output voltage and current are 3 ϕ AC. The peak value of the line to line voltage of each phase is 230 V. The peak value of output current is 5 A.

The output voltage is nearly sinusoidal but has a few ripples. The output voltage and current are lower than the input voltage and current. Hence, the output power is lower than the input power, which accounts for the losses.

4. Hardware Implementation

The experimental setup of the sparse AC link inverter is shown in Fig. 12. The gating pulses were created using a field programmable gate array (FPGA) board. The generated pulses are given to the switches of the multi mini capacitor (MMC) via the MCT2E opto-coupler. The power supply circuit for the optocoupler and processor was designed using a step-down transformer, bridge rectifier, and regulator.

A link inductor and link capacitor are present between the input and output sides to perform partial resonance and to provide galvanic isolation. In order to protect the microcontroller and the control circuit from power circuit an isolation circuit is provided. The experimental results are observed using digital storage oscilloscope (DSO). Figure 13 shows the DC link output voltage and current. The three-phase output voltage and current waveform from the inverter is shown in Fig. 14, where the voltage and current are in phase with each other. The synchronised current is shown for phase A.

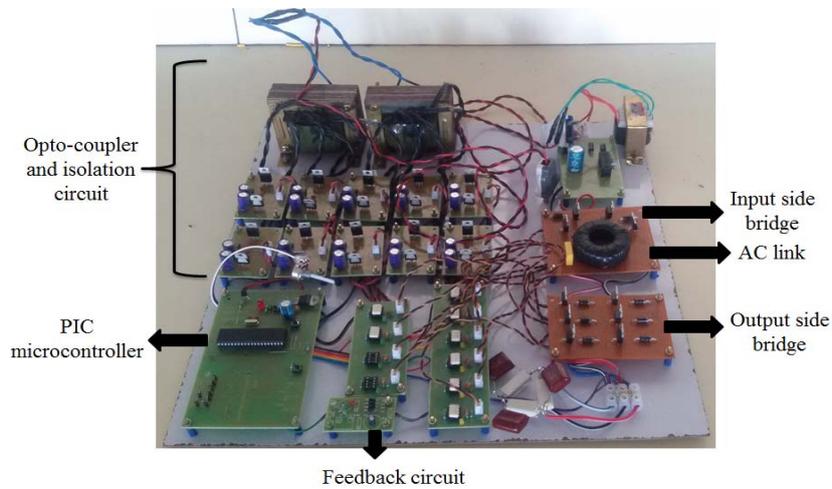


Figure 12. Hardware setup of the sparse AC-link inverter.

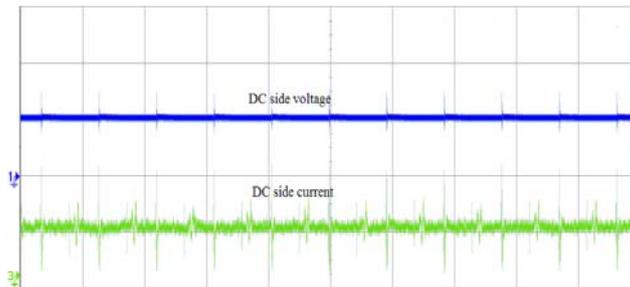


Figure 13. Direct current (DC) side voltage (200 V/division) and current (5 A/division).

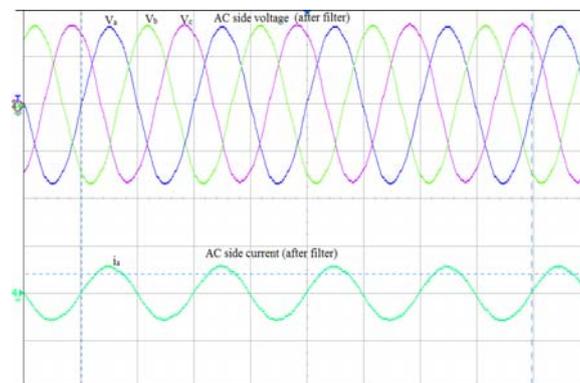


Figure 14. Alternating current (AC) side voltage (200 V/division) and current (5 A/division).

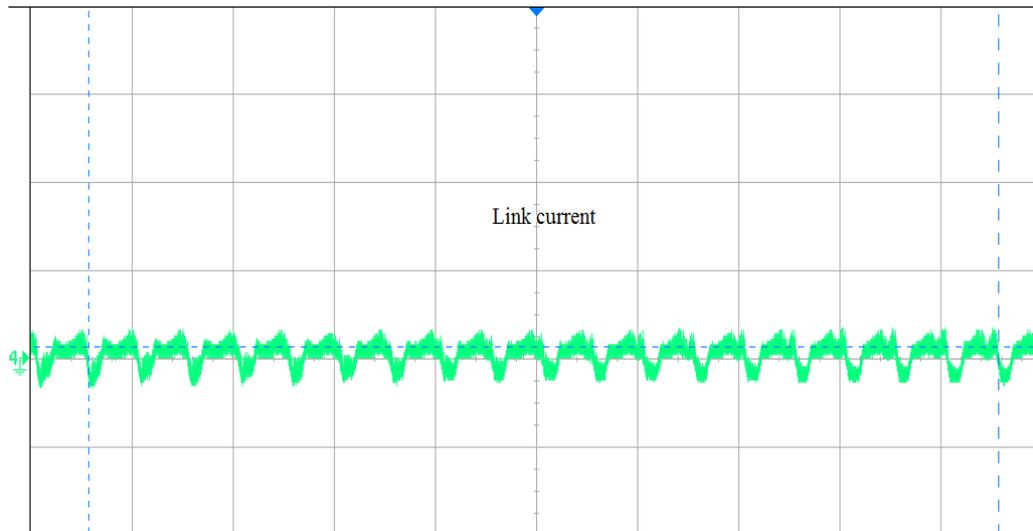


Figure 15. Link current (20 A/division).

The output AC link current was sinusoidal (Fig. 15). The output AC current obtained was approximately 5 A.

From the results, it can be observed that a nearly sinusoidal voltage and current were obtained using a sparse AC-link inverter with reduced switches.

5. Conclusion

This paper presents a soft-switched AC-link sparse inverter with reduced switches. The proposed inverter has some advantages such as zero voltage upon turn ON and soft turn OFF of switches. Although the proposed inverter has reduced switches compared to the existing sparse inverter, the efficiency of the system is not affected. The inverter has an LC (Inductor-Capacitor) pair which acts as a partial resonant circuit. This inverter provides isolation between the input and output without a transformer. Therefore, the inverter is more reliable and compact for a grid-connected PV system. The performance of the inverter was evaluated through both simulated and experimental results.

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